

# Streamline RF Synthesizer VCO Calibration and Optimize PLL Lock Time



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## ABSTRACT

When VCO-integrated RF synthesizers were first introduced, the typical implementation was a single VCO-core synthesizer. As a consequence, there were numerous devices with the same PLL but different VCO frequency cores to support different applications. Today's modern RF synthesizers are integrated with several VCO cores and output dividers such that a single device can support very low to very high output frequencies. To ensure all VCO cores work in harmony, proper VCO calibration is required. The calibration is critical to VCO frequency selection and meaningful to the frequency switching time. This application note addresses why VCO calibration is necessary, its theory of operation, and the methodology to minimize the calibration time for the LMX2594 wideband RF synthesizer. This application note applies to the LMX2594, LMX2595 and LMX2615-SP devices.

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## 1 Wideband VCO Implementation

An inductor in parallel with a capacitor forms a resonant tank circuit, as shown in [Figure 1-1](#).

If the user replaces the capacitor with a varactor diode and applies a different voltage to the varactor diode, the user can control the resonant frequency.

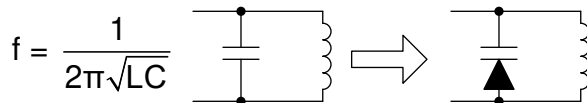


Figure 1-1. LC Resonant Tank Circuit

### 1.1 Discrete VCO

[Figure 1-2](#) shows a typical Colpitts oscillator that uses a varactor diode to realize frequency tuning. A tuning voltage,  $V_{tune}$ , is used to change the capacitance of the varactor diode, and therefore, change the frequency of the oscillator.

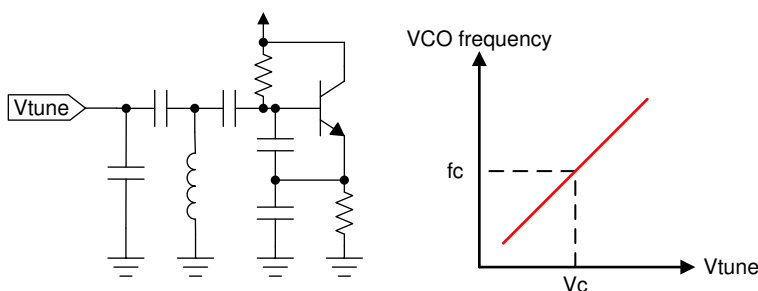


Figure 1-2. Voltage Control Oscillator

Assume when  $V_{tune} = V_c$ , the frequency of the VCO is  $f_c$ .

If the VCO gain (aka  $K_{vco}$ ) is 10 MHz/V, then the corresponding output frequency for  $V_c \pm 1$  V is  $f_c \pm 10$  MHz.

The frequency tuning range of a discrete VCO cannot be very wide. In general, the tuning range is about 10 to 15% of the VCO center frequency.

### 1.2 Silicon-Based, Switched-Capacitor VCO

If the VCO is built in a chip, it is possible to switch different capacitors in and out of the application to adjust the free running frequency of the VCO.

In other words, with  $V_{tune}$  being fixed, the center frequency of the VCO is adjustable. For example, if  $C_1$  is switched in, the corresponding output frequency is  $f_1$ . As a result, the frequency tuning range of the VCO has been extended from  $f_c \pm V_{tune}$  to  $f_1 \pm V_{tune}$ .

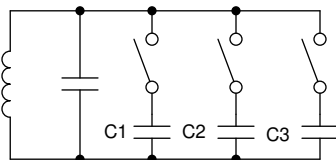


Figure 1-3. Switched-Capacitor Tank Circuit

### 1.3 VCO Capcode, VCO Core and VCO Amplitude

The term capcode in this document represents the value of the switched capacitors. In LMX2594, the capcode (register R19, VCO\_CAPCTRL) is between 0 and 255. The higher the capcode value, the lower is the VCO frequency. To support 7.5-GHz to 15-GHz VCO tuning range, a single switched-capacitor VCO is still not possible. The LMX2594 uses 7 VCO cores (register R20, VCO\_SEL) to cover this range. Each VCO has similar capcode structure but different inductor value.

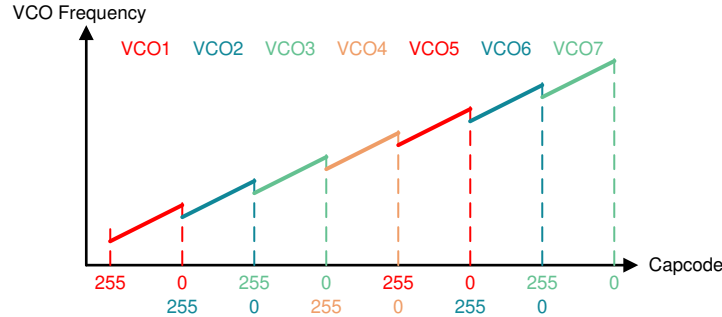


Figure 1-4. VCO Capcode

#### Note

Each adjacent VCO core has some overlapping regions. This is necessary to ensure seamless VCO core transition and to account for process-voltage-temperature (PVT) variations.

With 256 capcodes and 7 VCO cores, there are altogether 1792 combinations. For a particular frequency, it may be difficult to manually select the right VCO core and capcode, if not impossible. A VCO calibration algorithm is embedded in the chip so that the selection of capcode and VCO core is transparent to the users. In addition to frequency selection, the calibration algorithm also optimizes the phase noise by adjusting the bias of the selected VCO (register R16, VCO\_DACISSET).

## 2 VCO Calibration in LMX2594

Although the VCO calibration has been made as user-friendly as possible, take care to ensure that the calibration is robust and returns the optimum performance.

In the coming sections of this application note, the TICS Pro configuration as shown in Figure 2-1 is used as the default configuration for demonstration.

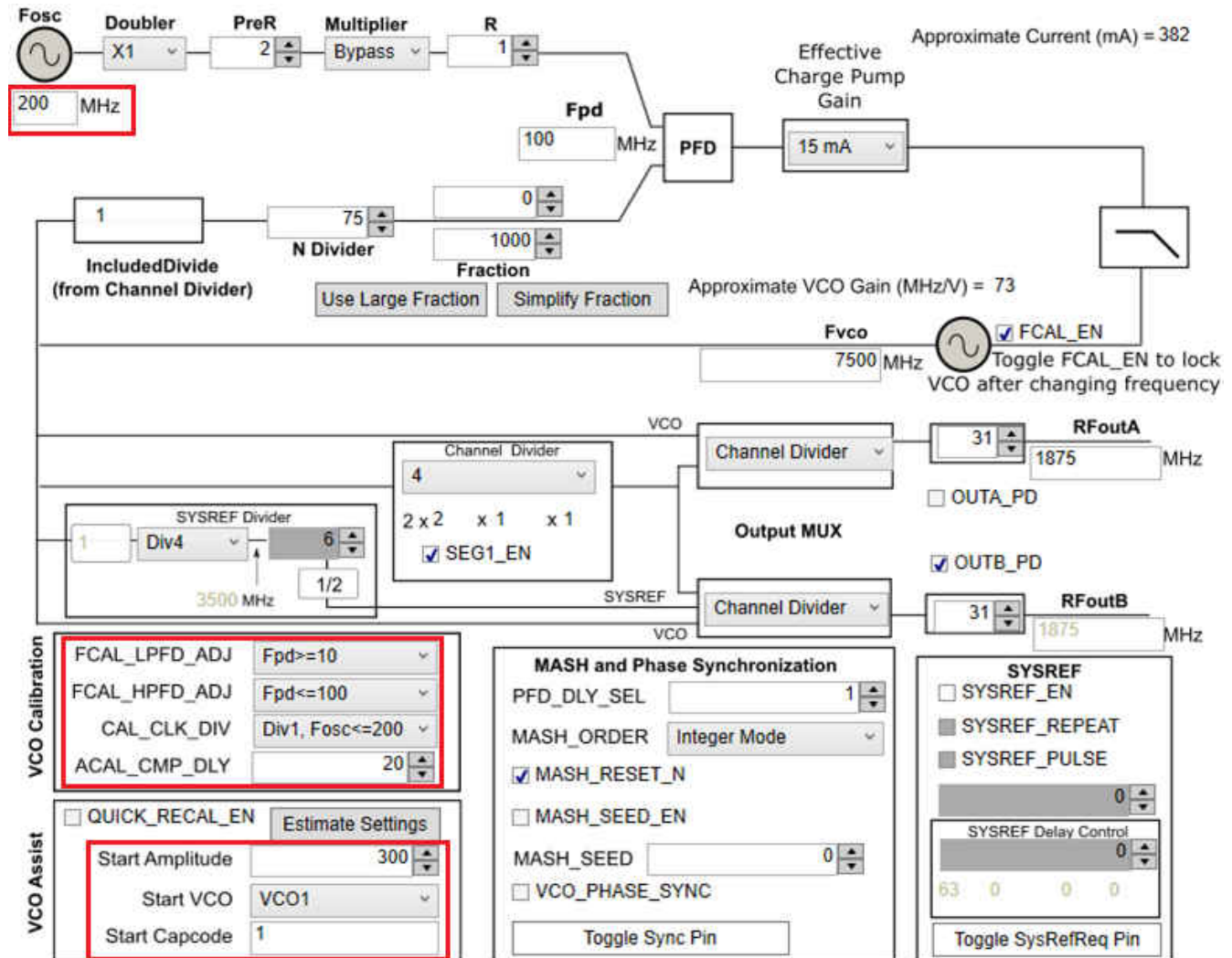


Figure 2-1. Default TICS Pro Configuration

Note the following with regards to the default configuration:

- To provide better illustration, Channel Divider was set to 4 so that the frequency scale is smaller and therefore easier to read.
- To simplify operation, only integer channels are used for demonstration.
- By default, the LMX2594 will scan the VCO core in ascending order.

### 2.1 Calibration Speed

The speed of calibration will affect the accuracy of a VCO calibration most. Of course, the speed will also impact on the calibration time directly. The highlighted items in Figure 2-1 should be set properly to balance between the calibration time and accuracy. How these parameters affect the calibration behavior is shown in the following sections.

### 2.1.1 State Machine Clock

The VCO calibration of the LMX2594 runs on state machine clock frequency,  $f_{SM}$ . The clock source of the state machine clock comes from the reference clock. Maximum  $f_{SM}$  is 200 MHz. Register R1, CAL\_CLK\_DIV, is used to set this frequency.

$$f_{SM} = f_{OSC} / 2^{CAL\_CLK\_DIV} \quad (1)$$

In the following plots, the VCO jumped from 7500 MHz to 15000 MHz with a different state machine clock frequency. Programming sequence is:

1. Program register R36, PLL\_N, from 75 to 150 (decimal).
2. Program register R0 once to trigger the VCO calibration.

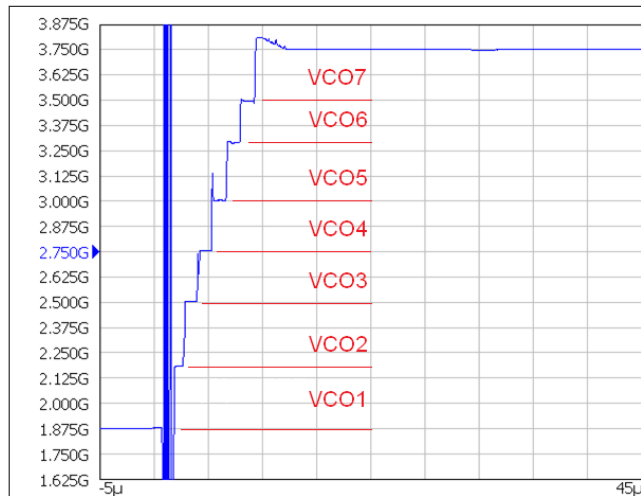


Figure 2-2.  $f_{SM} = 200$  MHz

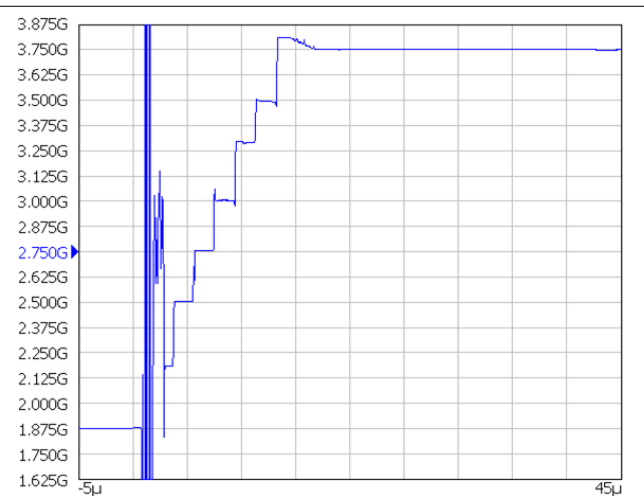


Figure 2-3.  $f_{SM} = 100$  MHz

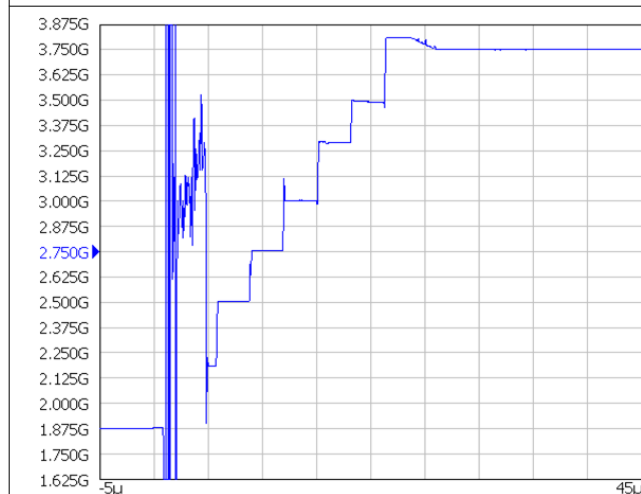


Figure 2-4.  $f_{SM} = 50$  MHz

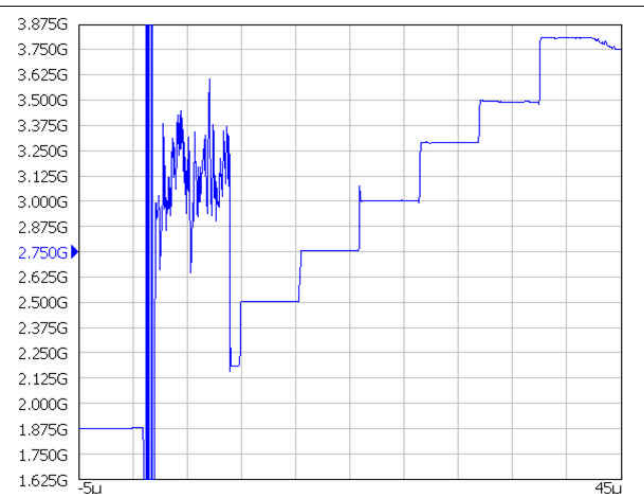


Figure 2-5.  $f_{SM} = 25$  MHz

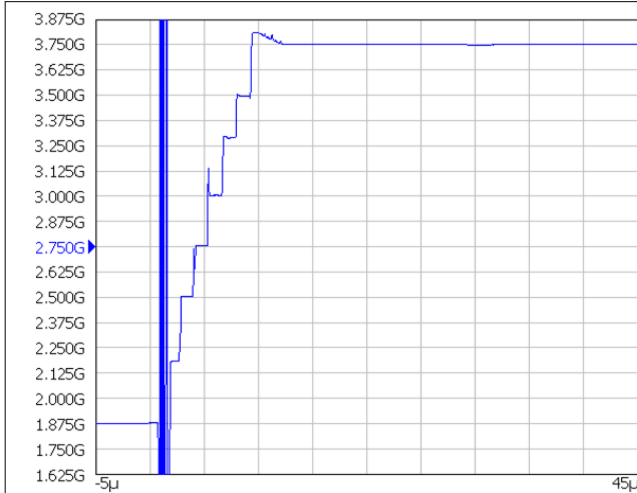
As shown above, the higher the  $f_{SM}$ , the shorter the calibration time. If the frequency switching time is important in certain applications, a higher reference clock frequency source is needed.

### 2.1.2 FCAL\_xxxx\_ADJ

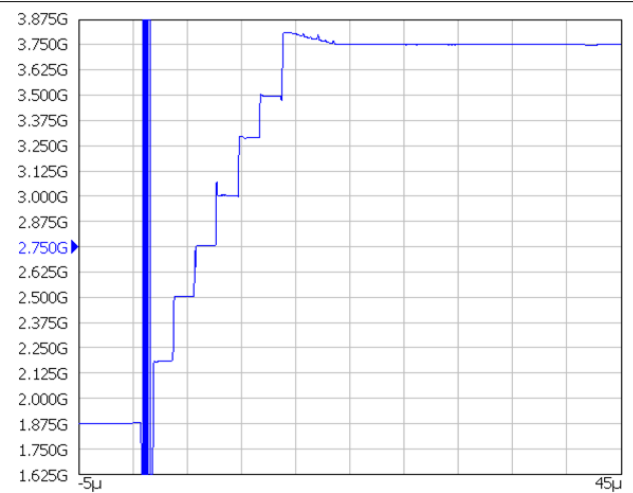
In Register R0, FCAL\_LPFD\_ADJ and FCAL\_HPFD\_ADJ are used to adjust the calibration speed when the phase detector frequency,  $f_{PD}$ , is too small or too high, respectively.

In most of the use cases,  $f_{PD}$  will not go below 10 MHz, so FCAL\_LPFD\_ADJ is basically not used at all.

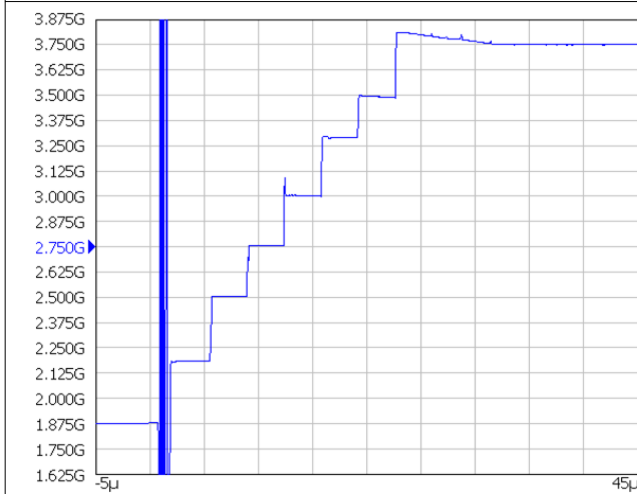
On the contrary, if  $f_{PD}$  is too high, FCAL\_HPFD\_ADJ should be set properly to slow down the calibration speed.



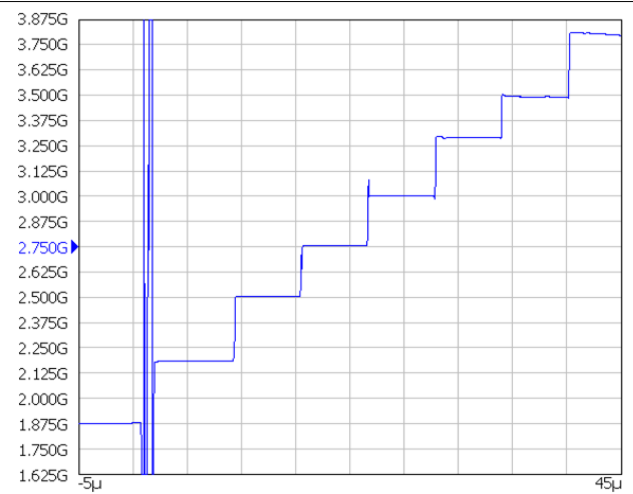
**Figure 2-6. FCAL\_HPFD\_ADJ = 0x0 ( $f_{PD} \leq 100$  MHz)**



**Figure 2-7. FCAL\_HPFD\_ADJ = 0x1 ( $100 < f_{PD} \leq 150$  MHz)**



**Figure 2-8. FCAL\_HPFD\_ADJ = 0x2 ( $150 < f_{PD} \leq 200$  MHz)**



**Figure 2-9. FCAL\_HPFD\_ADJ = 0x3 ( $f_{PD} > 200$  MHz)**

The state machine clock frequency remains unchanged here, but the calibration speed is reduced with a higher value of FCAL\_HPFD\_ADJ.

### 2.1.3 ACAL\_CMP\_DLY

In Register R4, ACAL\_CMP\_DLY is a delay added for VCO amplitude calibration. During amplitude calibration, the bias voltage of the VCO will be adjusted step-by-step until the best phase noise is obtained. This bias voltage comes from an internal LDO. It takes time for the LDO voltage to settle down, therefore a delay is necessary for every step voltage change. Lowering this value can speed up VCO calibration, but lowering it too much may degrade VCO phase noise. The minimum allowable value for this field is 10 to allow the VCO to calibrate to the correct frequency for all scenarios. To yield the best and most repeatable VCO phase noise, use [Equation 2](#) to find the ACAL\_CMP\_DLY value.

$$\text{ACAL\_CMP\_DLY} > f_{\text{SM}} / 10 \text{ MHz} \quad (2)$$

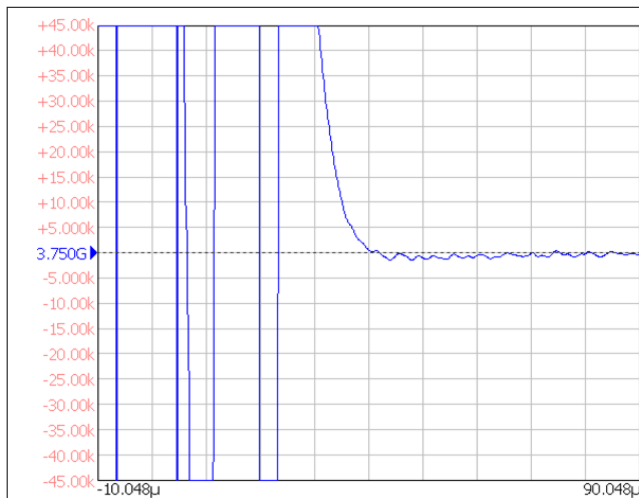


Figure 2-10. ACAL\_CMP\_DLY = 10, Jump Up

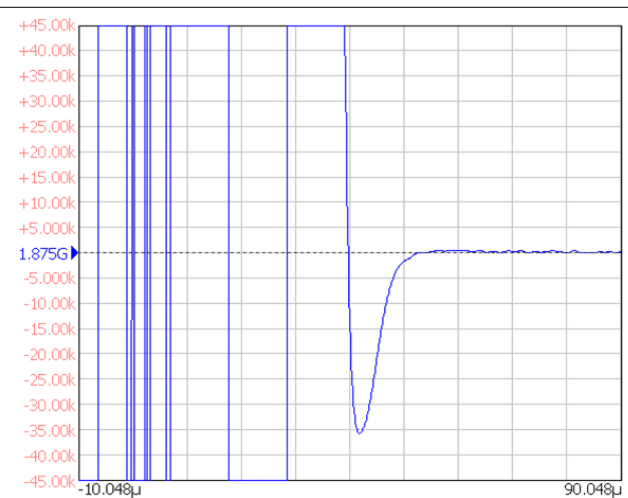


Figure 2-11. ACAL\_CMP\_DLY = 10, Jump Down

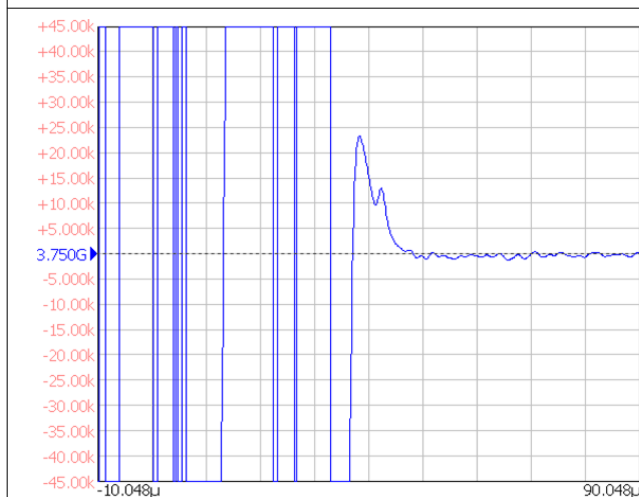


Figure 2-12. ACAL\_CMP\_DLY = 20, Jump Up

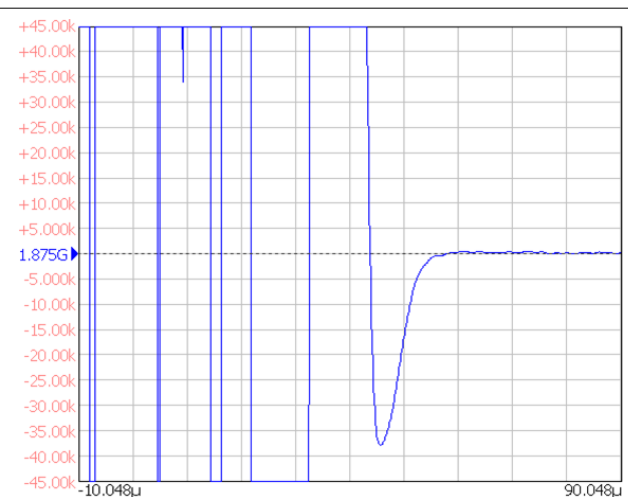
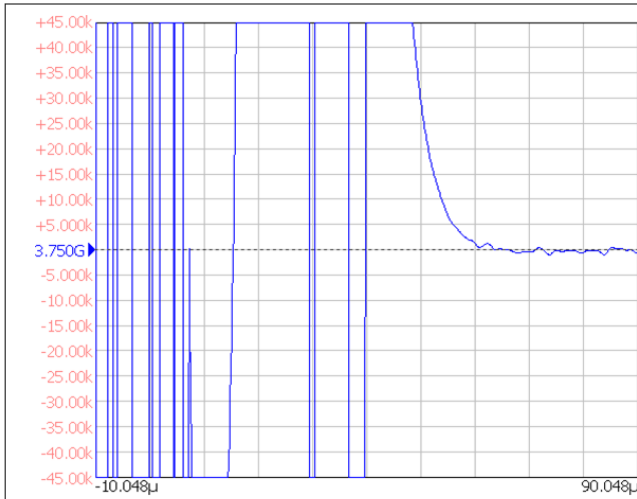
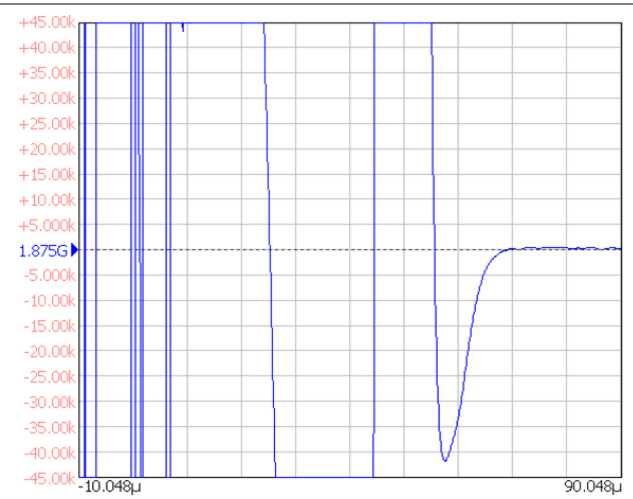


Figure 2-13. ACAL\_CMP\_DLY = 20, Jump Down





**Figure 2-14. ACAL\_CMP\_DLY = 40, Jump Up**



**Figure 2-15. ACAL\_CMP\_DLY = 40, Jump Down**

On the left column of the above plots, the VCO jumped from 7500 MHz to 15000 MHz. On the right column, the VCO frequency is flipped.

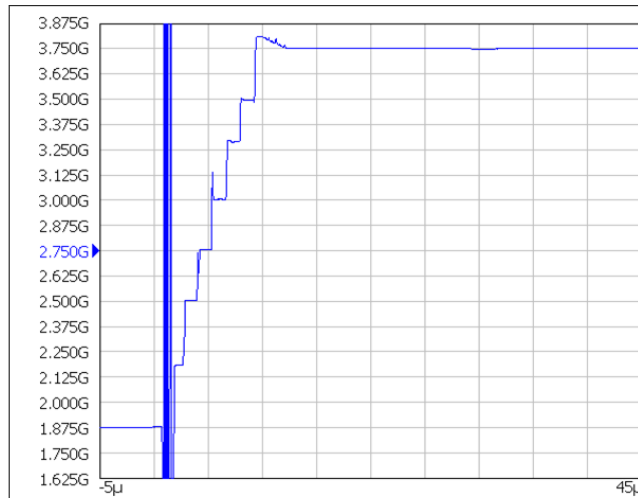
The state machine clock used in these examples is 200 MHz. Equation 2 suggests ACAL\_CMP\_DLY = 20. From the plots, the total lock time is increased by approximately 10  $\mu$ s, compared with the minimum ACAL\_CMP\_CAL requirement of 10.

## 2.2 Initial Calibration Parameters Setting

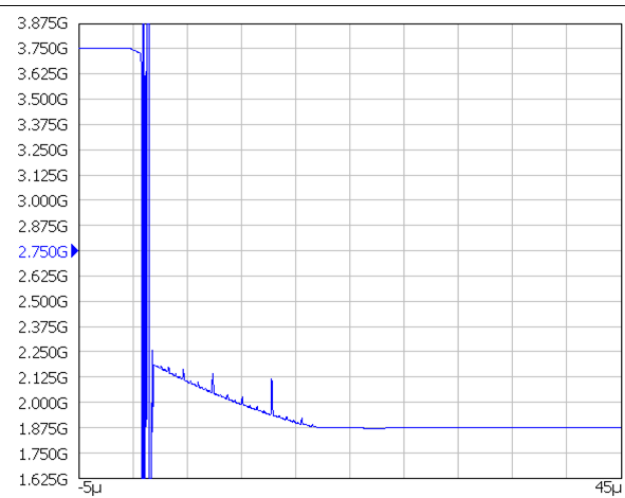
VCO calibration will always start at a pre-defined VCO core, VCO capcode and amplitude setting, no matter what is the present setting. These settings will affect the calibration time.

## 2.2.1 VCO\_SEL

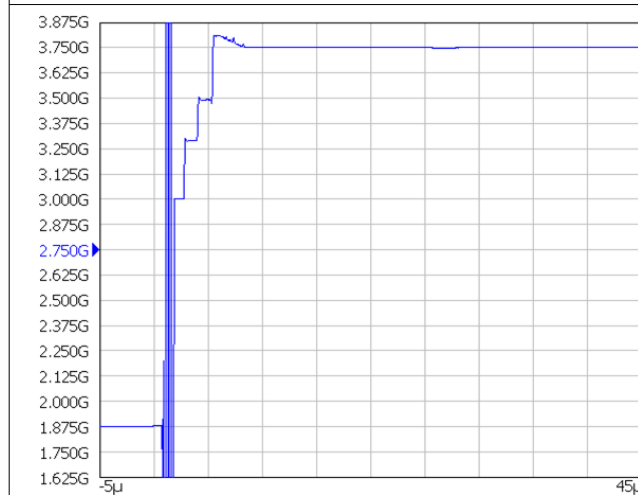
In Register R20, VCO\_SEL defines the VCO core that is being used at the beginning of a VCO calibration. The following examples compare the calibration time with different VCO\_SEL settings, with the VCO jumping upward and downward.



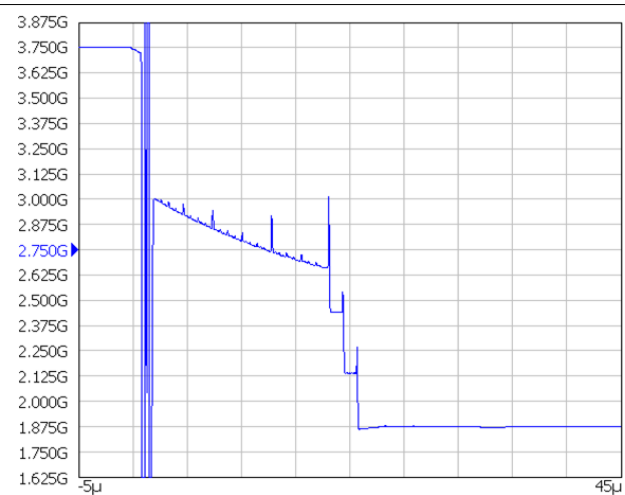
**Figure 2-16. VCO\_SEL = 1, Jump Up**



**Figure 2-17. VCO\_SEL = 1, Jump Down**



**Figure 2-18. VCO\_SEL = 4, Jump Up**



**Figure 2-19. VCO\_SEL = 4, Jump Down**

VCO\_SEL affects both jump up and jump down calibration time. In the above examples, jump up time is shorter if we use VCO4 as the starting VCO. However, the jump down time is longer than when VCO1 is being used as the starting VCO. There may be benefits to set it to either VCO1 or VCO7 in some use cases, a well-balanced approach should make it start at the middle of the total VCO coverage range.

## 2.2.2 VCO\_CAPCTRL\_STRT

Register R78, VCO\_CAPCTRL\_STRT sets the initial capcode value being used for VCO calibration. Capcode values are between 0 and 255. 0 represents the highest VCO frequency setting while 255 drives the VCO to the lowest frequency boundary.

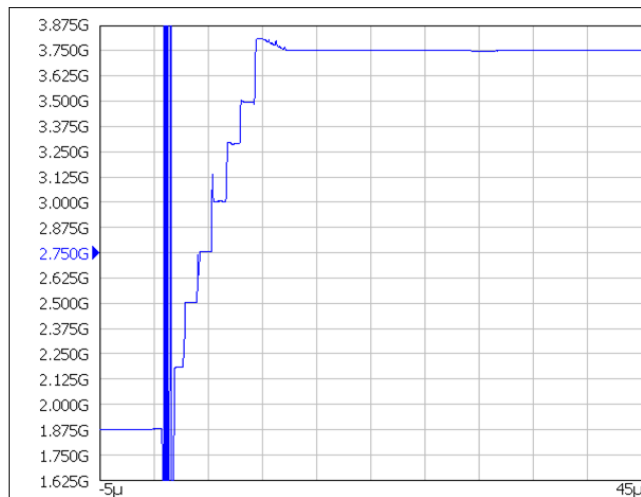


Figure 2-20. VCO\_CAPCTRL\_STRT = 1, Jump Up

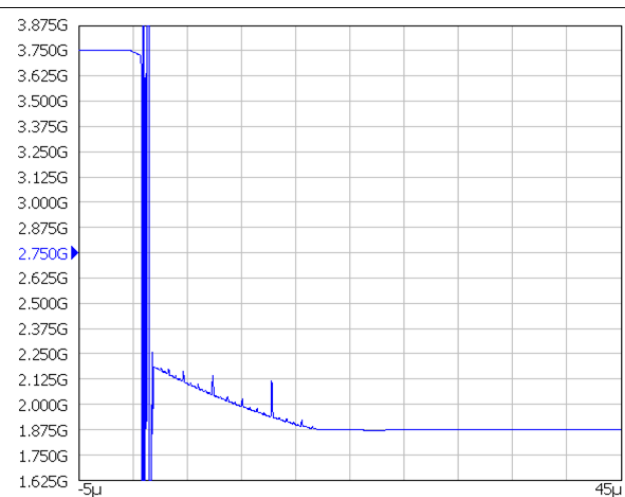


Figure 2-21. VCO\_CAPCTRL\_STRT = 1, Jump Down

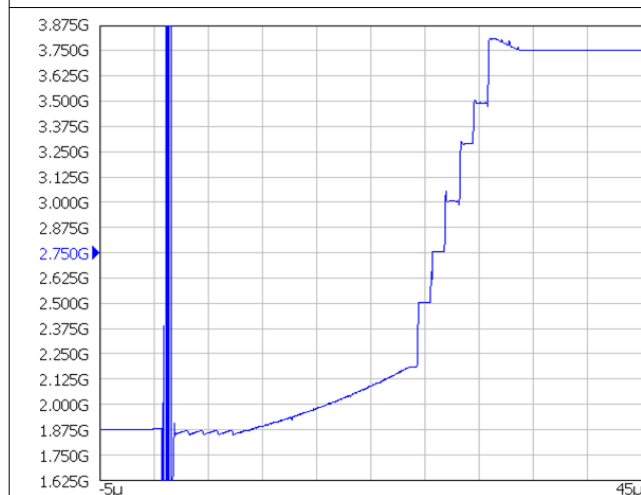


Figure 2-22. VCO\_CAPCTRL\_STRT = 255, Jump Up

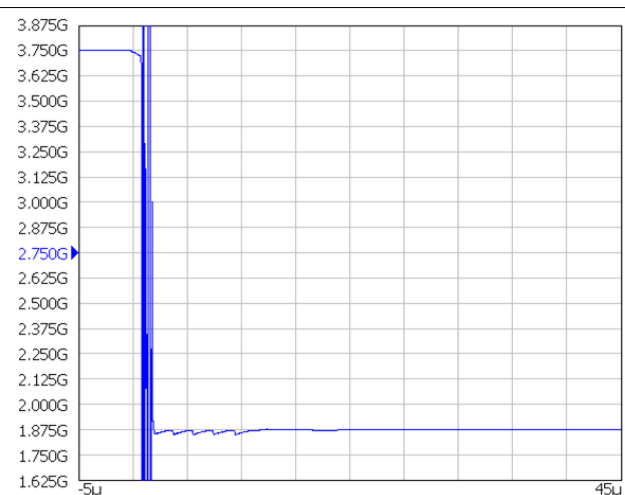


Figure 2-23. VCO\_CAPCTRL\_STRT = 255, Jump Down

Due to the behavior of the VCO calibration algorithm, the calibration time will be longer in general if the calibration starts with capcode = 255.

## 2.2.3 VCO\_DACISSET\_STRT

In Register R17, VCO\_DACISSET\_START sets the VCO amplitude value for use in the beginning of VCO calibration. This register has relatively less impact to the calibration time.

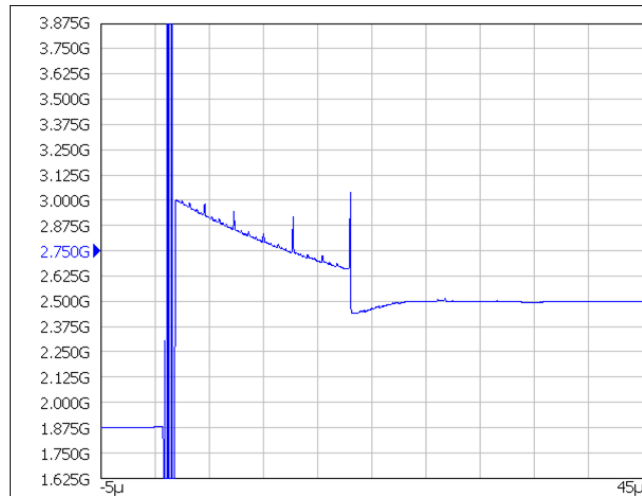
## 2.2.4 Recommended Initial Calibration Parameter Settings

To compromise jump up and jump down behaviors, the following initial calibration parameters are recommended:

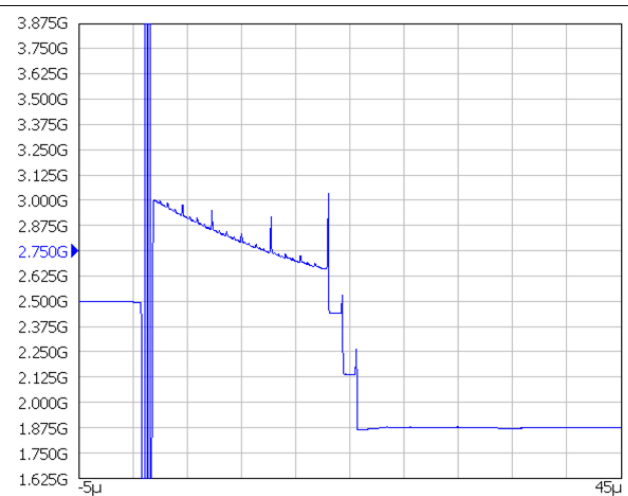
- VCO\_DACISSET\_STRT = 300
- VCO\_SEL = 4
- VCO\_CAPCTRL\_STRT = 1

With the suggested initial calibration parameters, here are some examples of how the VCO can jump up and down. The frequency of the VCO core in the following plots are:

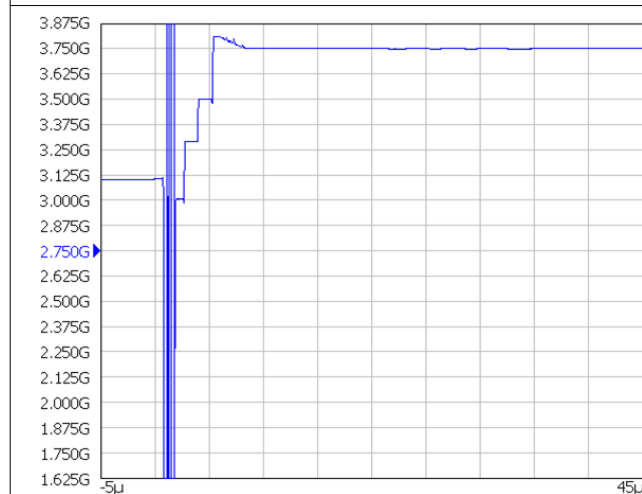
- VCO1 = 7500 MHz.
- VCO3 = 10 GHz.
- VCO5 = 12.4 GHz.
- VCO7 = 15 GHz.



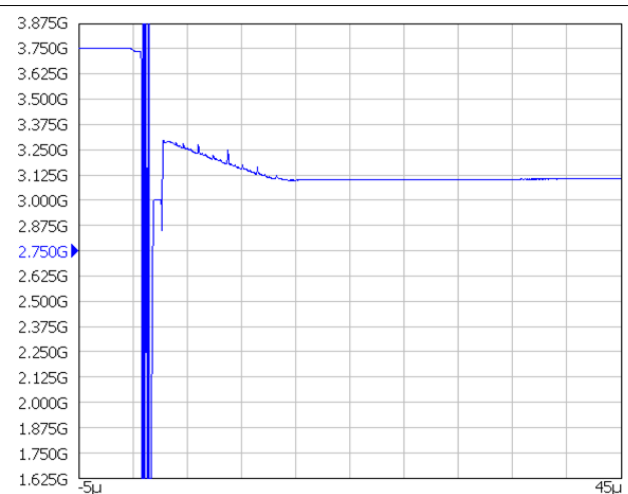
**Figure 2-24. VCO1 Jumps to VCO3**



**Figure 2-25. VCO3 Jumps to VCO1**



**Figure 2-26. VCO5 Jumps to VCO7**



**Figure 2-27. VCO7 Jumps to VCO5**

### 3 Reducing Calibration Time

Using the recommended initial calibration parameters, and with properly calibration speed setting, the VCO algorithm will pick the correct VCO core at the right capcode and amplitude setting. However, a sufficient amount of time must be reserved for the VCO calibration routine to complete. To have a control to the calibration time or wants to reduce the calibration time substantially, use partial assist or full assist mode in VCO calibration. With a shorter VCO calibration time, the total VCO frequency switching time will be smaller and therefore it is suitable for use in frequency hopping or sweeping system.

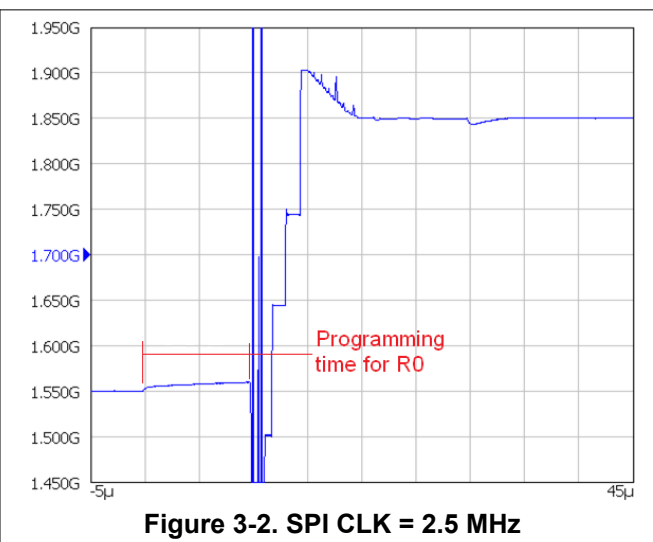
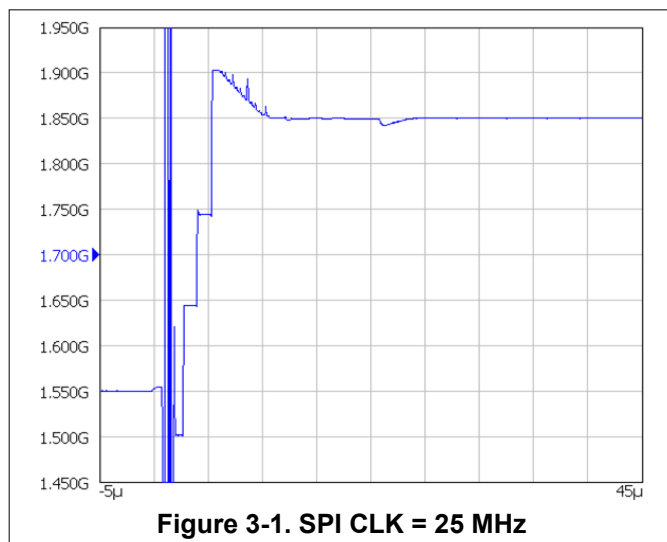
#### 3.1 SPI Programming Speed

In partial assist or full assist mode, in addition to the frequency related registers, such as N-divider and fractional numerator that must be updated, several additional registers need programming. Altogether, there are at least five registers that must be programmed. A register is 24 bits (DATA) + 1 bit (CSB) long, therefore, it will take 25  $\mu$ s to write a register for a 1-MHz SPI clock rate. Furthermore, the functions or values associated with a register are effective right after the register is written. If the N-divider is updated, for example, the VCO frequency drifts immediately and the PLL loop unlocks while the user is writing the R0 register. All of these changes take place before the VCO starts calibration.

The PLL loop is still closing, and the charge pump voltage goes to the rails as a result. This may affect the PLL lock time when the loop is closed again after the VCO calibration is complete, because the capacitors in the loop filter may take more time to charge up or discharge.

The following plots show the benefits of using a higher SPI clock rate. The VCO switched from VCO5 (12.4 GHz) to VCO7 (14.8 GHz) with the channel divider CHDIV = 8. The N-divider was updated first, followed by register R0, to trigger VCO calibration. As seen in Figure 3-2, even with a SPI clock rate of 2.5 MHz, a register write takes 10  $\mu$ s. The VCO, meanwhile, already drifted by approximately 80 MHz.

TI recommends to use a higher SPI clock rate whenever possible.



### 3.2 Partial Assist

In this mode, instead of using the initial calibration parameters as suggested in [Section 2.2.4](#), the user provides the initial starting point for the VCO core (VCO\_SEL), capcode (VCO\_CAPCTRL\_STRT), and amplitude (VCO\_DACISSET\_STRT) based on [Table 3-1](#) on every frequency change before writing to R0 to trigger VCO calibration. To do the partial assist, follow this procedure:

- Determine the VCO core.
  - Find a VCO core that includes the desired VCO frequency. If at the boundary between two cores, choose one based on phase noise or performance.
- Use [Equation 3](#) to calculate the VCO capcode.

$$\text{VCO\_CAPCTRL\_STRT} = \text{round} [C_{\text{CoreMin}} - (C_{\text{CoreMin}} - C_{\text{CoreMax}}) \times (f_{\text{VCO}} - f_{\text{CoreMin}}) / (f_{\text{CoreMax}} - f_{\text{CoreMin}})] \quad (3)$$

- Use [Equation 4](#) to calculate the VCO amplitude setting.

$$\text{VCO\_DACISSET\_STRT} = \text{round} [A_{\text{CoreMin}} - (A_{\text{CoreMin}} - A_{\text{CoreMax}}) \times (f_{\text{VCO}} - f_{\text{CoreMin}}) / (f_{\text{CoreMax}} - f_{\text{CoreMin}})] \quad (4)$$

**Table 3-1. LMX2594 VCO Core Ranges**

VCO CORE	f <sub>CoreMin</sub>	f <sub>CoreMax</sub>	C <sub>CoreMin</sub>	C <sub>CoreMax</sub>	A <sub>CoreMin</sub>	A <sub>CoreMax</sub>
VCO1	7500	8600	164	12	299	240
VCO2	8600	9800	165	16	356	247
VCO3	9800	10800	158	19	324	224
VCO4	10800	12000	140	0	383	244
VCO5	12000	12900	183	36	205	146
VCO6	12900	13900	155	6	242	163
VCO7	13900	15000	175	19	323	244

Here is an example.

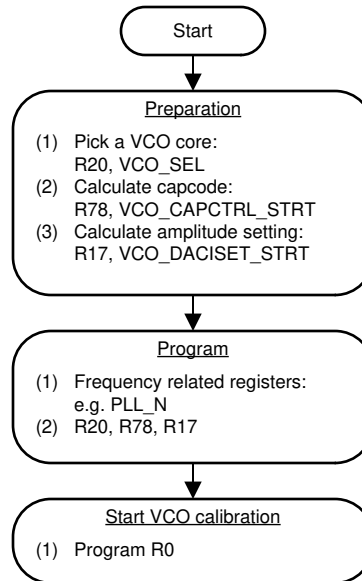
To jump from VCO1 (7500 MHz) to VCO3 (10 GHz), program the following registers:

- R36 = 0x240064 (PLL\_N = 100)
- R20 = 0x14D848 (VCO\_SEL = 3)
- R78 = 0x4E0105 (VCO\_CAPCTRL\_STRT = 130)
- R17 = 0x110130 (VCO\_DACISSET\_STRT = 304)
- R0 = 0x00241C (to trigger calibration)

To jump from VCO3 (10 GHz) to VCO1 (7500 MHz), program the following registers:

- R36 = 0x24004B (PLL\_N = 75)
- R20 = 0x14C848 (VCO\_SEL = 1)
- R78 = 0x4E0149 (VCO\_CAPCTRL\_STRT = 164)
- R17 = 0x11012B (VCO\_DACISSET\_STRT = 299)
- R0 = 0x00241C (to trigger calibration)

The test results are shown in [Figure 3-7](#) and [Figure 3-8](#).



**Figure 3-3. Partial Assist Workflow**

### 3.3 Full Assist

In this mode, VCO calibration is completely bypassed. The user forces the VCO core (VCO\_SEL), amplitude settings (register R16, VCO\_DACISSET), and capcode (register R19, VCO\_CAPCTRL), and manually sets the value. To get these values for a particular VCO frequency, the user must run an automatic calibration in advance. The user sets an automatic calibration for this frequency, then reads back these values in register R110, rb\_VCO\_SEL; R111, rb\_VCO\_CAPCTRL and R112, rb\_VCO\_DACISSET. To use these values, the user must enable register R20, VCO\_SEL\_FORCE as well as the R8, VCO\_CAPCTRL\_FORCE and VCO\_DACISSET\_FORCE bits.

Here is an example, assuming VCO\_SEL\_FORCE, VCO\_CAPCTRL\_FORCE and VCO\_DACISSET\_FORCE are all equal 1.

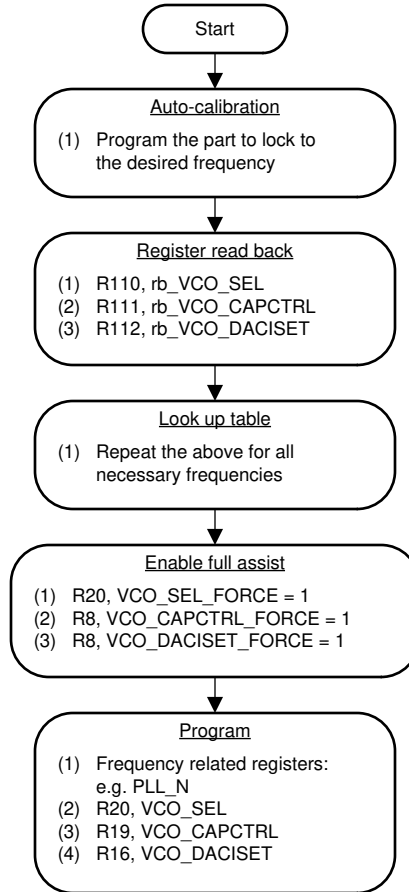
To jump from VCO1 (7500 MHz) to VCO3 (10 GHz), program the following registers:

- R36 = 0x240064 (N = 100)
- R20 = 0x14DC48 (VCO\_SEL = 3)
- R19 = 0x132787 (VCO\_CAPCTRL = 135)
- R16 = 0x10012E (VCO\_DACISSET = 302)

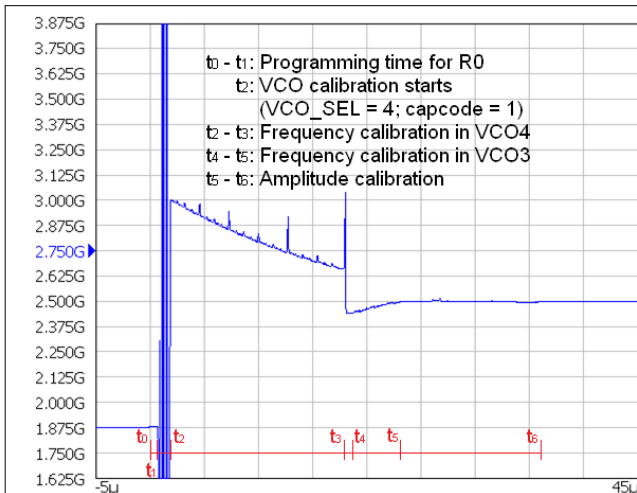
To jump from VCO3 (10 GHz) to VCO1 (7500 MHz), program the following registers:

- R36 = 0x24004B (N = 75)
- R20 = 0x14CC48 (VCO\_SEL = 1)
- R19 = 0x1327AB (VCO\_CAPCTRL = 171)
- R16 = 0x10012C (VCO\_DACISSET = 300)

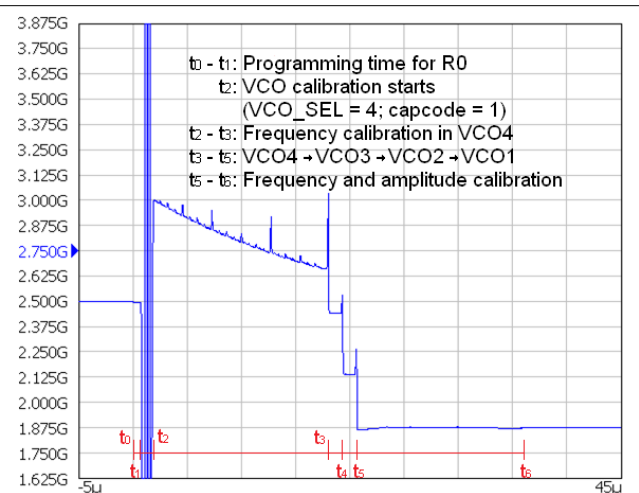
The test results are shown in [Figure 3-9](#) and [Figure 3-10](#).



**Figure 3-4. Full Assist Workflow**

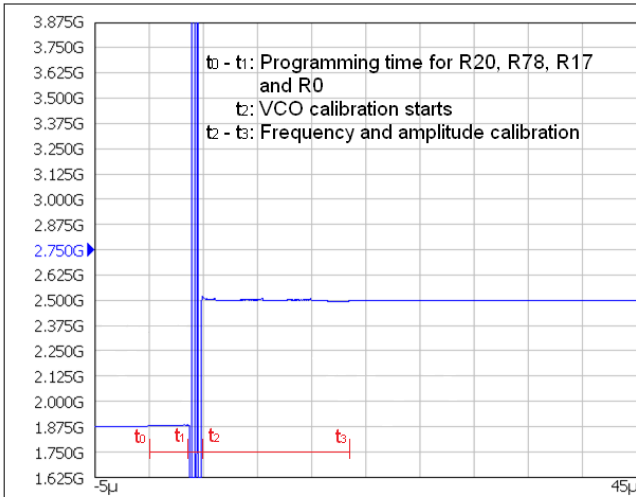


**Figure 3-5. Auto-Calibration, Jump Up**

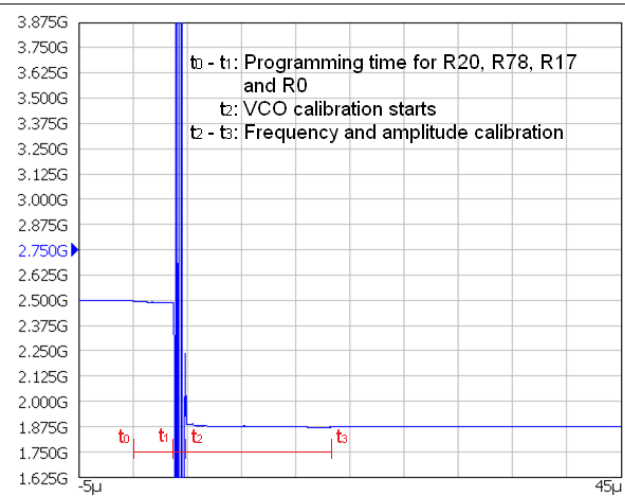


**Figure 3-6. Auto-Calibration, Jump Down**

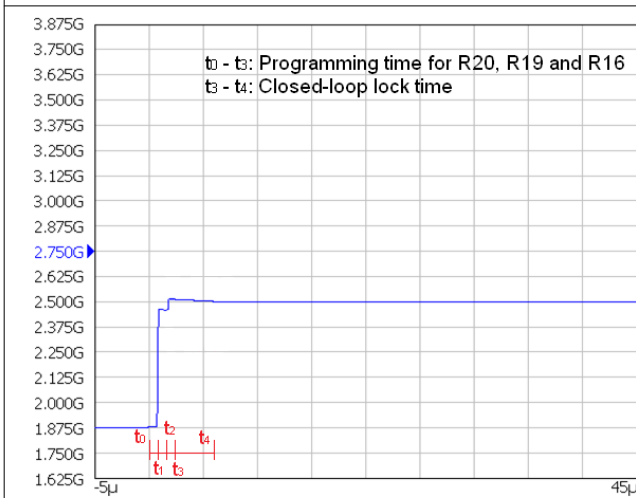




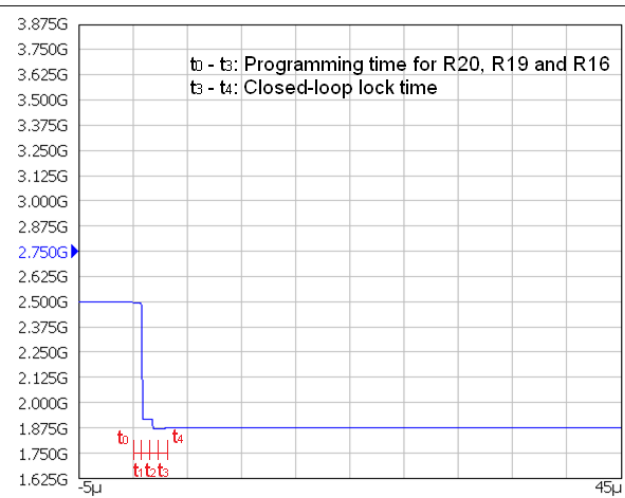
**Figure 3-7. Partial Assist, Jump Up**



**Figure 3-8. Partial Assist, Jump Down**



**Figure 3-9. Full Assist, Jump Up**



**Figure 3-10. Full Assist, Jump Down**

## 4 Summary

With auto-calibration, the calibration time varies but without hassle. To reduce the calibration time, the user can use the listed methods in this document to provide specific calibration parameters. SPI programming speed is also important and should be set fast enough to reduce the unlock duration when the VCO is drifting before the calibration starts.

## 5 References

Related products information and tools are available in the following links:

- [LMX2594](#)
- [LMX2595](#)
- [LMX2615-SP](#)
- [TICS Pro software](#)
- [PLLatinum Simulator Tool \(PLL Sim\)](#)

## 6 Revision History

NOTE: Page numbers for previous revisions may differ from page numbers in the current version.

Changes from Revision * (January 2020) to Revision A (July 2021)	Page
• Updated the numbering format for tables, figures and cross-references throughout the document.....	3

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